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Correction of High-Frequency Noise-Temperature Inaccuracies

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Deep-space mission data rates to Earth are limited by the system operating noise-temperature (T_{op}) performance of the DSN. This article addresses some of the techniques and definitions used for measuring and reporting the effective noise temperature of receivers (T_e) and T_{op} of the DSN's ground receiving systems. Calibration loads are used to measure T_{op} of the DSN antennas. At 32 GHz, a calibration load cooled to 2-K physical temperature requires a correction of 0.67 K to determine the noise temperature. Using corrected noise temperature for the calibration loads results in the correct values for T_{op} such that the total system noise power can be defined by $P_n = kT_{op} B$, as required for DSN telecommunications design control tables. T_{op} and T_e should not be converted to equivalent physical temperatures.

I. Introduction

System operating noise temperature (T_{op}) is very important to the DSN; deep-space mission data rates to Earth are limited by the DSN's T_{op} performance. The DSN uses design control tables to document parameters of the spacecraft-to-ground end-to-end telecommunications system. A key parameter affecting the data quality is the signal-to-noise ratio (SNR) of the signal received by the DSN. The received SNR is proportional to DSN antenna gain divided by the system operating noise temperature (G/T_{op}).

This article addresses some of the techniques and definitions used for measuring and reporting the effective

noise temperature of receivers (T_e) and T_{op} of the DSN's ground receiving systems. Proper evaluation of the noise-temperature performance of high-frequency, low-noise amplifiers (LNA's), such as Ka-band (32-GHz) masers currently under development in the DSN Advanced Systems Program, requires the use of frequency-dependent corrections for the noise power available from the calibration loads. At 32 GHz, a calibration load cooled to 2 K has an available noise power equivalent to 1.33 K; a correction of 0.67 K is needed.

An analysis for optimizing the testing configuration for LNA noise temperature is provided in [1]; frequency-dependent corrections are not used. Frequency-dependent

[4] can give large errors. Using Eq. (2) with the equivalent corrected noise temperature results in

$$P_n = kT_n B \quad (4)$$

Use of the corrected equivalent noise temperature is appropriate for telecommunications design control tables, such as used in the DSN [5].

Calibration of the corrected equivalent noise temperature of an LNA using a load with physical T and a noise source (usually a noise diode connected to the LNA through a directional coupler between the LNA and the load) with temperature TND , all referred to the amplifier input, requires solution of

$$T_e = \frac{TND}{Y-1} - T_n \quad (5)$$

where

Y = power ratio at the output of follow-up amplifiers with the noise source turned on and off

TND = noise source excess noise at amplifier input, K

T_n = equivalent noise temperature of source at physical temperature T , K

Measurement configurations using cooled attenuators located between the load and the LNA are evaluated using Eq. (5) by analyzing an equivalent TND and T_n defined at the LNA input. Similarly, using two calibration loads and a cooled attenuator requires the evaluation of corrected equivalent noise temperatures T_1 and T_2 for the loads, defined at the LNA input, and the solution of

$$T_e = \frac{T_2 - T_1 Y}{Y - 1} \quad (6)$$

where Y = power ratio obtained at output of follow-up amplifiers switching between T_1 and T_2 .

The value T_e , as measured with Eqs. (5) and (6), contains the follow-up amplifier noise temperature Tf . The LNA noise temperature, $TLNA$, requires the correction

$$TLNA = T_e - Tf \quad (7)$$

For most system applications, especially when the first amplifier has more than a 30-dB gain, Tf is small compared

with T_e . For system applications, T_e is the significant parameter.

III. Results

Equations (5) and (6) have been programmed in Supercalc 4. Assuming the load and noise source are separated from the amplifier by an attenuator with loss L at physical temperature T_p , Figs. 1 and 2 show the solutions using Eq. (5). Using Eq. (6), Figs. 3 and 4 assume the loads are separated from the amplifier by the attenuator. The values TND , L , T_p , T_1 , T_2 , and Y are considered known. Equation (3) is used to correct T_1 , T_2 , and TL for frequency. The equation $TL = T_p (1 - 1/L)$ represents the attenuator noise-temperature contribution. The Eq. (3) correction is applied to T_p , not TL .

For the purposes of this article, $f = 0.001$ GHz is used as the dc ($f = 0$) case. The results shown in Fig. 1, at dc, assume $Y = 2.1136$, appropriate for $T_e = 4.0$ K and the other input parameters assumed and used in [2]. Figure 2 shows the result of operating at $f = 32$ GHz with all other inputs unchanged. The errors in T_e due to various parameter changes are virtually unchanged with frequency and also agree with [2] (Fig. 3 for $L = 20$ dB). However, T_e increases from 4.00 to 4.67 K at $f = 32$ GHz relative to dc. Figures 3 and 4 have similar results, with T_e increasing from 4.00 K at dc to 4.68 K at 32 GHz.

IV. Conclusion

Using loads with corrected equivalent noise temperatures results in the proper value for the amplifier noise temperature, T_e . The value T_e in this case is the equivalent noise temperature, not the physical temperature. From Eq. (3), the physical temperature is given by

$$T = \frac{hf/k}{\ln((hf/kT_n) + 1)} \quad (8)$$

Equation (8) is useful for converting a measured noise temperature, T_n , to a physical temperature, T . An example of this is using T , the physical or thermodynamic temperature, for reporting the cosmic background radiation temperature. The cosmic physical or thermodynamic temperature obtained with Eq. (8) after measuring the noise temperature is independent of frequency.

The physical or thermodynamic temperature is not appropriate for reporting measurements of amplifier noise

temperature, T_e , for such purposes as the DSN telecommunications-link design tables. In addition, quantum noise is inherent to low-noise amplifiers and is included in

at S- and X-bands presently used in the DSN (T reduced by 0.2 K at 8.4 GHz for a 300-K load) but is important for future Ka-band operation (T reduced by 0.77 K at 32 GHz).

Table 1. Examples of errors in treating T_c as a constant with changes of frequency, f , and the calibration load physical temperature, T .

f , GHz	T , K	T_c , K	Error in T_c , K due to 10-percent change in T
8	2	0.19	0.0006
8	80	0.19	<0.0001
8	300	0.19	<0.0001
32	2	0.67	0.0087
32	80	0.77	0.0002
32	300	0.77	<0.0001

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INPUT -----
      T= 300   TND= 1000   Tp= 2   f,GHz= .001
      DT= .1   DTND= 50   DTp= .1   Y= 2.1136
      DYLDL,A= .01   DLDB,A= .01   B,MHZ= 50   DYG= .01
      DYLDL,B= .01   DLDB,B= .03   T,SEC= 1   L= 100

RESULTS -----"Te error (DTe)-----
      DL .78622          DTp .09900          DYG .328408
      DT .00100          DYL .16455          SUM 1.83299
      DTND .44899        DYN .00482          RMS .982087

CALCULATIONS -----
      L,DB= 20   Y,DB= 3.2502   DYN= .00014   hf/k= .000048
      NOMINAL          DELTA CALCULATIONS -----
                        L+DL,DB=      T+DT=      TN+DTN
                        20.61          300.1          1050
                        L+DL=
                        115.08
      TL= 1.98          1.9826          1.98          1.98
      TLn= 1.9800      1.9826          1.9800          1.97998
      TnR= 3.0000      2.6069          3.0010          3.00000
      T= 4.9800        4.5895          4.9810          4.97998
      TND= 10          8.6896          10          10.5
      Te= 3.9999      3.2137          3.9989          4.44890

DELTA CALCULATIONS,CONT -----
      Tp+DTp=          Y+DYL,DB=          Y(1+2*DYN)=          Y(1+2*DYG)=
      2.1              3.2927              2.1142              2.15587
                        Y+DY=
                        2.1344
      TL= 2.079          1.98          1.98          1.98
      TLn= 2.0790      1.9800          1.9800          1.97998
      TnR= 3.0000      3.0000          3.0000          3.00000
      T= 5.0790        4.9800          4.9800          4.97998
      TND= 10          10          10          10
      Te= 3.9009      3.8354          3.9951          3.67150

DEFINITIONS
Te=LNA NOISE TEMP          Tp=PHY TEMP OF L
L=ATTEN LOSS              DTp=DELTA Tp
DL=DELTA L                TND=NOISE DIODE
TL=TEMP CONTR OF L        DTND=DELTA NOISE DIODE
Y=Y FACTOR                 T=LOAD TEMP
DYL=DELTA Y FACTOR,LINERITY DT=DELTA T
DYN=DELTA Y, RADIOMETER NOISE (T,B) Tn=T CORRECTED FOR FREQ
DYG=DELTA Y, RADIOMETER GAIN DELTA G TnR=Tn AT REF

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Fig. 1. Supercalc 4 computer program NOISE2ND printout of the measured noise temperature and errors of a low-noise amplifier using a load, noise diode, and cooled attenuator at 0.001 GHz (dc).

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INPUT -----
T= 300 TND= 1000 Tp= 2 f,GHz= 32
DT= .1 DTND= 50 Dtp= .1 Y= 2.1136
DYLDB,A= .01 DLDB,A= .01 B,MHZ= 50 DYG= .01
DYLDB,B= .01 DLDB,B= .03 T,SEC= 1 L= 100

RESULTS -----"Te error (DTe) -----
DL .78635 Dtp .10794 DYG .328408
DT .00100 DYL .16455 SUM 1.84206
DTND .44899 DYN .00482 RMS .983130

CALCULATIONS -----
L,DB= 20 Y,DB= 3.2502 DYN= .00014 hf/k= 1.536
NOMINAL DELTA CALCULATIONS -----
L+DL,DB= T+DT= TN+DTN
20.61 300.1 1050
L+DL=
115.08
Tpn= 1.3294 1.3294 1.3294 1.32935
TLn= 1.3161 1.3178 1.3161 1.31606
TnR= 2.9923 2.6002 2.9933 2.99233
T= 4.3084 3.9180 4.3094 4.30838
TND= 10 8.6896 10 10.5
Te= 4.6715 3.8852 4.6705 5.12050

DELTA CALCULATIONS,CONT -----
Tp+Dtp= Y+DYL,DB= Y(1+2*DYN)= Y(1+2*DYG)=
2.1 3.2927 2.1142 2.15587
Y+DY=
2.1344
Tpn= 1.4384 1.3294 1.3294 1.32935
TLn= 1.4240 1.3161 1.3161 1.31606
TnR= 2.9923 2.9923 2.9923 2.99233
T= 4.4163 4.3084 4.3084 4.30838
TND= 10 10 10 10
Te= 4.5636 4.5070 4.6667 4.34309

DEFINITIONS
Te=LNA NOISE TEMP Tpn=Tp CORRECTED FOR FREQ
L=ATTEN LOSS Tp=PHY TEMP OF L
DL=DELTA L Dtp=DELTA Tp
TL=TEMP CONTR OF L TND=NOISE DIODE
Y=Y FACTOR DTND=DELTA NOISE DIODE
DYL=DELTA Y FACTOR,LINEARITY T=LOAD TEMP
DYN=DELTA Y, RADIOMETER NOISE (T,B) DT=DELTA T
DYG=DELTA Y, RADIOMETER GAIN DELTA G Tn=T CORRECTED FOR FREQ
TnR=Tn AT REF

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Fig. 2. Supercalc 4 computer program NOISE2ND printout of the measured noise temperature and errors of a low-noise amplifier using a load, noise diode, and cooled attenuator at 32 GHz.

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INPUT -----
      T2= 300      T1= 80      Tp= 2      f,Ghz= .001
      DT2= .1      DT1= 1      DTP= .01      Y= 2.5942
      DYLDB,A= .01 DLDB,A= .01 B,MHZ= 50      DYG= .01
      DYLDB,B= .01 DLDB,B= .03 T,SEC= 1      L= 10

RESULTS ----- TE error (DTE) -----
      DL .41335      DTP .00900      DYG .434972
      DT2 .00627      DYL .26228      SUM 1.29495
      DT1 .16273      DYN .00635      RMS .674902

CALCULATIONS -----
      L,DB= 10      Y,DB= 4.1400      DYN= .00014      hf/k= .000048
      NOMINAL      DELTA CALCULATIONS -----
                        L+DL,DB=      T2+DT2=      T1+DT1=
                        10.31      300.1      81
                        L+DL=
                        10.740
      TL= 1.8      1.8138      1.8      1.8
      TLn= 1.8000      1.8138      1.8000      1.79998
      T2nR= 30.000      27.933      30.010      30.0000
      T2 31.800      29.747      31.810      31.8000
      T1nR= 8.0000      7.4489      8.0000      8.10000
      T1 9.8000      9.2626      9.8000      9.89997
      Te= 4.0001      3.5867      4.0063      3.83732

DELTA CALCULATIONS,CONT -----
      Tp+DTP=      Y+DYL,DB=      Y(1+2*DYN)=      Y(1+2*DYG)=
      2.01      4.1914      2.5949      2.64608
      Y+DY=
      2.6251
      TL= 1.809      1.8      1.8      1.8
      TLn= 1.8090      1.8000      1.8000      1.79998
      T2nR= 30.000      30.000      30.000      30.0000
      T2 31.809      31.800      31.800      31.8000
      T1nR= 8.0000      8.0000      8.0000      8.00000
      T1 9.8090      9.8000      9.8000      9.79997
      Te= 3.9911      3.7378      3.9937      3.56508

DEFINITIONS
Te=LNA NOISE TEMP      Tp=PHY TEMP OF L
L=ATTEN LOSS      DTP=DELTA Tp
DL=DELTA L      T1=COLD LOAD TEMP
TL=TEMP CONTR OF L      DT1=DELTA T1
Y=Y FACTOR      T2=HOT LOAD TEMP
DYL=DELTA Y FACTOR,LINEARITY      DT2=DELTA T2
DYN=DELTA Y, RADIOMETER NOISE (T,B)      Tn=T CORRECTED FOR FREQ
DYG=DELTA Y, RADIOMETER GAIN DELTA G      TnR=Tn AT REF

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Fig. 3. Supercalc 4 computer program NOISE2LD printout of the measured noise temperature and errors of a low-noise amplifier using two loads and a cooled attenuator at 0.001 GHz (dc).

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INPUT -----
      T2= 300      T1= 80      Tp= 2      f,Ghz= 32
      DT2= .1      DT1= 1      DTP= .01      Y= 2.5942
      DYLD, A= .01  DLDB, A= .01  B, MHZ= 50      DYG= .01
      DYLD, B= .01  DLDB, B= .03  T, SEC= 1      L= 10

RESULTS ----- TE error (DTE) -----
      DL .41400      DTP .00857      DYG .434968
      DT2 .00627      DYLD .26228      SUM 1.29516
      DT1 .16272      DYN .00635      RMS .675288

CALCULATIONS -----
      L, DB= 10      Y, DB= 4.1400      DYN= .00014      hf/k= 1.536
      NOMINAL      DELTA CALCULATIONS -----
                        L+DL, DB=      T2+DT2=      T1+DT1=
                        10.31      300.1      81
                        L+DL=
                        10.740
      Tpn= 1.3294      1.3294      1.3294      1.32935
      Tln= 1.1964      1.2056      1.1964      1.19642
      T2nR= 29.923      27.862      29.933      29.9233
      T2= 31.120      29.067      31.130      31.1197
      T1nR= 7.9234      7.3776      7.9234      8.02344
      T1= 9.1199      8.5832      9.1199      9.21986
      Te= 4.6801      4.2660      4.6863      4.51733

DELTA CALCULATIONS, CONT -----
      Tp+DTP=      Y+DYLD, DB=      Y(1+2*DYN)=      Y(1+2*DYG)=
      2.01      4.1914      2.5949      2.64608
      Y+DY=
      2.6251
      Tpn= 1.3389      1.3294      1.3294      1.32935
      Tln= 1.2050      1.1964      1.1964      1.19642
      T2nR= 29.923      29.923      29.923      29.9233
      T2= 31.128      31.120      31.120      31.1197
      T1nR= 7.9234      7.9234      7.9234      7.92345
      T1= 9.1284      9.1199      9.1199      9.11986
      Te= 4.6715      4.4178      4.6737      4.24508

DEFINITIONS
Te=LNA NOISE TEMP
L=ATTEN LOSS
DL=DELTA L
TL=TEMP CONTR OF L
Y=Y FACTOR
DYLD=DELTA Y FACTOR, LINEARITY
DYN=DELTA Y, RADIOMETER NOISE (T,B)
DYG=DELTA Y, RADIOMETER GAIN DELTA G

Tpn=Tp CORRECTED FOR FREQ
Tp=PHY TEMP OF L
DTP=DELTA Tp
T1=COLD LOAD TEMP
DT1=DELTA T1
T2=HOT LOAD TEMP
DT2=DELTA T2
Tn=T CORRECTED FOR FREQ
TnR=Tn AT REF

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Fig. 4. Supercalc 4 computer program NOISE2LD printout of the measured noise temperature and errors of a low-noise amplifier using two loads and a cooled attenuator at 32 GHz.